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Pravin Varaiya (M '68), for a photograph and biography, see this issue, p. 128.

# Teams, Signaling, and Information Theory

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*Abstract*—The purpose of this paper is to unify results from three separate and, at least superficially, unrelated subject matters, namely, team decision theory, market signaling in economics, and the classical Shannon information theory.

## I. INTRODUCTION

THE STUDY of the role of different information in many person decision problems called team theory was initiated by Marschak in the 1950's. More recently, through the pioneer work of Witsenhausen and others, this has been extended and unified with literatures on decentralized or nonclassical stochastic control theory which emphasized the role of information structure in problems involving dynamics, or sequential order of ac-

tions. During the same period sporadic and not too successful attempts have been made to relate Shannon's information theory with feedback control system design. Again with the recent maturity of control theory as a subject in applied mathematics, the two disciplines begin to exhibit much closer connection than heretofore displayed, e.g., the Viterbi algorithm and the Kalman-Bucy filter, the recent work of Whittle and Rudge [9]. Lastly, one of the current interests in mathematical economics is associated with the role of information in organizations and the market place. Various interesting phenomena arise as a result of imperfect or incomplete information in person-to-person interactions. The purpose of this paper is to attempt to weave a common thread among these three apparently unrelated subjects; team theory, market signaling, and information theory. While no particularly significant new results are obtained, we believe the conceptual unity displayed here among the three subjects is new and, hopefully, will lead to much future cooperative effort among researchers in these different fields. We should point out that the first hint of the connection between dynamic team theory and information-theoretic issues was due to Witsenhausen [4, p. 146] and [7, p. 334] (see problem (P-1) and (A) in Section III later). Also, the idea of signaling in a noneconomic context has also been used previously in game [24] and stochastic control theory [25].

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## II. FUNDAMENTALS OF INFORMATION STRUCTURE AND DECENTRALIZED DECISION MAKING

There are five basic ingredients of decision theory.

(1) The *state of the world*  $\xi \equiv [\xi_1, \dots, \xi_n] \in \Omega$  which can be thought of as a vector of random variables defined on a probability space having a density (or distribution or measure)  $p(\xi)$ .  $\xi$  represents all the uncertainties in the problem under consideration, e.g., unknown initial conditions, measurement noise, uncertain parameters, etc.

(2) A set of *decision variables*  $u \equiv [u_1, \dots, u_m] \in U$  each representing one decision maker (DM). One person making two decisions at different times is regarded as two DMs in this setup.

(3) A *loss (payoff) function* which is a measurable function of  $u$  and  $\xi$ , i.e.,  $L(u, \xi)$ . We assume  $L$  is expressed in appropriate utility units.

(4) A set of *information functions*  $z \equiv \eta(\xi) \in Z \equiv [\eta_1(\xi), \dots, \eta_m(\xi)]$ , one for each DM. In other words  $z_i = \eta_i(\xi)$ , what  $DM_i$  knows, is in general different from  $z_j$ , what  $DM_j$  knows. Alternatively, in place of  $\eta(\xi)$ , we can be given subalgebras induced by the  $\eta$ 's on the underlying probability space. The set of  $\eta$ 's or the subalgebras are known as the *information structure* of the problem.

(5) A set of *strategies*  $\gamma \equiv [\gamma_1, \dots, \gamma_m] \in \Gamma$ , one for each DM, where  $\gamma_i$  is a mapping from the  $z_i$ -space to the  $u_i$ -space. Thus, each DM must choose actions  $u_i = \gamma_i(z_i)$  based on different information. It is in this sense the problem is decentralized.

Since for fixed  $\gamma$ ,  $E[L(u = \gamma(\eta(\xi)), \xi)]$  is well-defined<sup>1</sup> and depends on  $\gamma$ , we can state the decision problem as

$$\text{Min}_{\gamma \in \Gamma} J(\gamma) = \text{Min}_{\gamma \in \Gamma} E[L(u = \gamma(\eta(\xi)), \xi)]$$

a deterministic optimization problem in the  $\Gamma$ -space which is usually taken to be the space of all measurable functions from  $U \equiv \prod_{i=1}^m U_i$  to  $Z \equiv \prod_{i=1}^m Z_i$ .

This problem is known as the *static team problem*. It is static in the sense that information  $z_i$  available to DM depends only on  $\xi$ . The evaluation of posterior probability such as  $p(\xi/z_i)$  can be separately carried out from the problem of choosing the actions  $u_i$ . However, in general when different DM's act at different times, information  $z_i$  received later by  $DM_i$  may be dependent upon the action  $u_j$  of  $DM_j$ , who acted earlier. Thus, in general decision problems, we must consider

$$(4') \quad z = \eta(\xi, u) \equiv [\eta_1(\xi, u), \dots, \eta_m(\xi, u)]$$

where  $\eta$  must satisfy some causality conditions [1]. When the team problem is characterized by the information structure (4') instead of (4), it is called a *dynamic team problem* [5]. The word dynamic is used to indicate the presence of order of actions of the DM's.

A superficially simple example which we shall use throughout this paper is now stated below.

Let  $\xi = [x, v]$  where  $x \sim N(0, 1)$  and  $v \sim N(0, \sigma^2)$ ,  $x, v$  independent.

$$(6) \quad \begin{aligned} L(u, \xi) &= \frac{1}{2}(x + au_1 + bu_2)^2 + \frac{1}{2}cu_1^2 \\ a, b, & \quad c \geq 0 \\ z_1 &= x \\ z_2 &= gx + hu_1 + v \quad g, h \geq 0, h = ga. \end{aligned}$$

One interpretation of the example is that  $x$  is the initial condition; the state after  $DM_1$  acts is  $x_1 \equiv x + au_1$ ; similarly  $x_2 \equiv x_1 + bu_2$ ;  $z_1$  is the measurement of the initial state by  $DM_1$ , and  $z_2$  is a noisy measurement of a linear transformation of  $x_1 = x + au_1$  by  $DM_2$ . The objective is to minimize the final state  $x_2$  and the energy, or power,  $\frac{1}{2}cu_1^2$  of  $DM_1$ , a control-theoretic problem. Note that this information structure is dynamic and that  $DM_1$  can signal or control the knowledge of  $x$  to  $DM_2$  through this action  $u_1 = \gamma_1(x)$ . A rather different interpretation can be given if we take  $a = g = 0$ ,  $b = -1$ ,  $h = 1$ . In this case,  $DM_1$ , knowing  $x$ , is trying to transmit a decision  $u_1$ , subject to energy constraints, through a noisy media so that  $DM_2$  can act based on  $z_2$  in order to minimize the difference (distortion) between  $x$  and  $u_2$ . If  $DM_1$  is called the "encoder" and  $DM_2$  the "decoder", then the information-theoretic significance of this interpretation is obvious [7]. In any case, this example appears to be the simplest type of team decision problem which incorporates dynamic information structure<sup>2</sup> [2] and all its attendant complexities.

With regard to the general team problem, the conditions for optimality are (letting  $m = 2$  for simplicity)

$$(P-1) \quad \begin{cases} \text{find } \gamma_1^*, \gamma_2^* \ni \\ J(\gamma_1^*, \gamma_2^*) \leq J(\gamma_1, \gamma_2) \quad \forall \gamma_1, \gamma_2 \in \Gamma. \end{cases}$$

A necessary condition for  $\gamma_1^*$  and  $\gamma_2^*$  to satisfy (P-1) is that they solve

$$(P-2) \quad \begin{cases} \text{find } \gamma_1^*, \gamma_2^* \ni \\ J(\gamma_1^*, \gamma_2^*) \leq J(\gamma_1, \gamma_2^*) \quad \forall \gamma_1 \in \Gamma \\ J(\gamma_1^*, \gamma_2^*) \leq J(\gamma_1^*, \gamma_2) \quad \forall \gamma_2 \in \Gamma. \end{cases}$$

which is known as person-by-person optimality (PBPO) or equilibrium solutions. The reason for the latter terminology becomes clear if we realize that, in general,  $DM_1$  need not necessarily have the same loss function or criterion of performance as  $DM_2$ . For  $i = 1, 2$ , let  $J_i$  be the criterion of  $DM_i$ , and let  $J_1 \neq J_2$ . No conceptual difficulties are involved if we extend (P-2) to

$$(P-2)' \quad \begin{cases} \text{find } \gamma_1^*, \gamma_2^* \ni \\ J_1(\gamma_1^*, \gamma_2^*) \leq J_1(\gamma_1, \gamma_2^*) \quad \forall \gamma_1 \in \Gamma \\ J_2(\gamma_1^*, \gamma_2^*) \leq J_2(\gamma_1^*, \gamma_2) \quad \forall \gamma_2 \in \Gamma. \end{cases}$$

This is known as Nash equilibrium in the parlance of

<sup>1</sup>Provided, of course,  $\gamma$  and  $\eta$  are appropriately measurable functions.

<sup>2</sup>Dynamic in the sense of (4').

game theory. If  $J_1 \neq J_2$ , the problem is called a nonzero-sum (Nzs) game. If  $J_1 = -J_2 \triangleq J$ , the problem is a zero-sum (Zs) game because  $J_1 + J_2 = 0$ . (P-2)' becomes

$$(P-3) \begin{cases} \text{find } (\gamma_1^*, \gamma_2^*) \ni \\ J(\gamma_1^*, \gamma_2) \leq J(\gamma_1^*, \gamma_2^*) \leq J(\gamma_1, \gamma_2^*) \end{cases}$$

the saddle point condition. With this condition, the example problem now admits a game-theoretic interpretation.  $DM_1$  wishes to act to cancel out  $x$  without using too much energy, but his action reveals the knowledge of  $x$  to  $DM_2$  through  $z_2$ :  $DM_2$  wishes to maximize the terminal state  $x + au_1 + bu_2$  which he can do if he knows  $x$  well.<sup>3</sup>

More will be said about (P-2)' and (P-3) later on in Section IV and elsewhere [3]. For the moment let us return to (P-1) and (P-2). The principal difficulties introduced by dynamic information structure (I)' are two-fold:

i) The observation  $z_2$  is not a well-defined random variable until the strategy  $\gamma_1$  is specified. This makes the various probability measures required in the solution process solution-dependent. There is a vicious circle and the problem of estimation is no longer separable from that of control.

ii) The optimization problem  $\text{Min}_{\gamma_1, \gamma_2} J(\gamma_1, \gamma_2)$  is not necessarily convex in  $\gamma_1$ . This is because  $\gamma_1$  enters in  $J(\gamma_1, \gamma_2(\gamma_1))$  also through  $\gamma_2(z_2) = \gamma_2(gx + h\gamma_1(x) + v)$ . Since there is no reason to expect  $\gamma_2$  to be convex, there is no assurance that  $J$  is convex in  $\gamma_1$  even though  $L$  may be convex in  $u_1$ .

Both difficulties were fully investigated by Witsenhausen [4] for the case of (P-1) with  $a=g=h=1$ . Since his seminal work, other efforts have been made to isolate cases where these difficulties can be circumvented [5]. In fact, it can be argued that whatever success we have in optimal stochastic control theory is based on the existence result under the special information structure of perfect memory which bypasses the above mentioned difficulties [6, p. 461].

### III. SIGNALING AND INFORMATION THEORY

In view of the difficulties mentioned above and in [4], it is somewhat surprising that, in fact, something can be done for (P-1). Suppose we take instead the information-theoretic interpretation of the problem (2) with  $a=g=0$ ,  $b=-1$ ,  $h=1$  (the problem in [4] is the same except that  $a=g=1$ ), but with average signal power constrained to be less than or equal to 1 (that is, choose  $c$  in (2) ap-

propriately). Then the problem becomes

$$(A) \begin{cases} \text{Min}_{\gamma_1, \gamma_2} E[(x - \gamma_2(\gamma_1(x) + v))^2] \\ \text{s.t. } E[\gamma_1(x)]^2 \leq 1 \\ \text{with } z_1 = x \\ z_2 = u_1 + v \end{cases}$$

with optimal solutions  $\gamma_1^*(x) = x$  and  $\gamma_2^*(z_2) = E(x/z_2) = 1/(1 + \sigma^2)z_2$ . But (A) is recognized as a special case of Shannon's information-theoretic problem involving a memoryless Gaussian source and an additive memoryless Gaussian channel where the source rate and channel rate are equal.<sup>4</sup> In the language of (A), this means we require the dimensions of  $x$  and  $u_1$ , if regarded as a vector, to be equal (and in the case of (A), equal to 1). The encoder  $\gamma_1(z_1)$ , and the decoder  $\gamma_2(z_2)$ , are *instantaneous and linear*. Here is a situation where information theory provided a solution to a dynamic team problem which, in the absence of this knowledge, would have been most difficult to solve.<sup>5</sup>

This interesting connection between team and information theory can be further exploited by considering several variants of (A). First, consider the general case with dimension  $(x) = n$  and  $\text{dim}(u_1) = m$ . In other words, the source has block length  $n$  and the channel block length  $m$ . Then (A) becomes

$$(A)' \begin{cases} \text{Min}_{\gamma_1, \gamma_2} \frac{1}{n} E[(x - \gamma_2(\gamma_1(x) + v))^T (x - \gamma_2(\gamma_1(x) + v))] \\ \text{s.t. } \frac{1}{m} E[\gamma_1(x)^T \gamma_1(x)] \leq 1 \\ z_1 = x = \begin{bmatrix} x_1 \\ \vdots \\ x_n \end{bmatrix} \triangleq \begin{bmatrix} z_{11} \\ \vdots \\ z_{1n} \end{bmatrix}, \\ x_i \sim N(0, 1), x_i \text{ and } x_j \text{ independent for } i \neq j \\ z_2 = u_1 + v = \begin{bmatrix} u_{11} + v_1 \\ \vdots \\ u_{1m} + v_m \end{bmatrix} \triangleq \begin{bmatrix} z_{21} \\ \vdots \\ z_{2m} \end{bmatrix}, \\ v_i \sim N(0, \sigma^2), v_i \text{ and } v_j \text{ independent for } i \neq j. \end{cases}$$

The variants of (A)' that we will consider are as follows:

$$(A-1) \begin{cases} \text{general } n \text{ and } m \\ \text{let } n \rightarrow \infty, \text{ but } \frac{n}{m} = \text{constant} \end{cases}$$

<sup>3</sup>Of course, some penalty term on  $u_2$  must be added to (2). See problem (C) later.

<sup>4</sup>See introduction of this paper and also Whittle and Rudge [9].  
<sup>5</sup>There exist no general sufficiency conditions to verify the optimality of a solution besides the Shannon bounds.

$$(A-2) \quad n = m$$

$$(A-3) \quad n = 1, m = 2$$

$$(A-4) \quad n = 2, m = 1$$

$$(A-5) \quad \begin{cases} n = 1, m = 2, \text{ but } z_1 = \begin{bmatrix} x \\ z_{21} \end{bmatrix} \\ \text{then } u_{11} = \gamma_{11}(x) \\ \quad u_{12} = \gamma_{12}(x, z_{21}). \end{cases}$$

Problem (A-1) is a statement of the well-known Shannon information theory problem [8, p. 911] where one is only concerned with the rate of information transmission ( $n/m = \text{constant}$ ) but initial delay is acceptable ( $n \rightarrow \infty$ ). The optimum performance (minimum distortion)  $J^*$  is known and is obtained when one equates  $R_{\text{eq}}(\beta)$ , the equivalent source rate for a given distortion  $\beta$ , to the equivalent channel rate  $C_{\text{eq}}(\alpha)$  for a given signal power level  $\alpha$  [8]. That is,  $J^* = \beta^*$ , where  $\beta^*$  is defined by  $R(\beta^*) = C_{\text{eq}}(\alpha)$ . However, the encoder and decoder ( $\gamma_1^*, \gamma_2^*$ ) pair to realize  $J^*$  is still unknown. Problem (A-1) can be generalized considerably by allowing memory in both the source and the channel. But even with memory, within the linear (memory structure)-quadratic (distortion)-Gaussian (source and channel) setup (LQG), the minimum distortion can still be obtained from Shannon theory (see Whittle and Rudge [9]).

Once we leave problem (A-1), it may be argued that we have entered the realm of "real-time information theory." In problems (A-2) through (A-5), we are not allowed to encode a large number of messages ( $x$ 's) together before transmission. Arbitrary delay is not permitted. Equivalently, the block length is fixed. The emphasis here is more decision-theoretic. However, much information-theoretic insight can still be borrowed to provide solutions or partial solutions to these problems, as the following discussion will show.

As mentioned earlier, Problem (A-2) corresponds to the situation where the channel rate and source rate are equal. The asymptotic results are the same as if the vector, or block, lengths are fixed: the optimal encoder and decoder are linear [9], [7], [21]. This is not true if  $n \neq m$ , as will now be discussed.

Problem (A-3) is the prototype of situations where one is allowed to signal more than once for each piece of information he wishes to send. In the language of communication, we are allowed to trade bandwidth for performance. Both of the following are optimal linear strategies:  $u_1 = \begin{bmatrix} x \\ x \end{bmatrix}$ , and  $u_1 = \begin{bmatrix} \sqrt{2} x \\ 0 \end{bmatrix}$ . The latter clearly shows that the only gain is in increasing power and not in making use of the expanded bandwidth. Hence, far better nonlinear strategies must exist, and a construction of a near-optimal strategy in the small noise case can be obtained by using Shannon's twisted modulation idea [22], [23]. It should be pointed out that the optimum  $J^*$  is not even known in this case, although a lower bound is possible via the Shannon theory. Whether or not a better bound is possible with a

different definition of mutual information in the spirit of Ziv and Zakai [10] is an open question.

Problem (A-4) is the opposite of (A-3) and is representative of source coding, where data compression is desired. In many respects, it is similar to a problem in optimal quantization. From topological considerations, we know that if the mapping is to be invertible in the absence of noise, then it cannot be continuous. Hence, even without considering the effect of noise in detail, we know that an optimal mapping must be nonlinear.<sup>6</sup> Similar remarks on  $J^*$  apply here as in (A-3).

Problem (A-5) is the same as (A-3) except that noiseless feedback is allowed.  $DM_1$  is allowed to send the second signal based on  $x$  and  $z_{21}$ . It turns out that the solution to this problem is known. The best  $(\gamma_1^*, \gamma_2^*)$  is linear for (A-5) and realizes the Shannon bound in real time [11], [12], [13], [26]. The solution consists of sending  $x$  as the first signal, then sending an amplified innovation term  $x - E(x/z_{21}) \equiv v$  which is independent of  $z_{21}$  as the second signal, resulting in  $z_{22} = kv + v_2$ , where  $k = ((1 + \sigma^2)/2)^{1/2}$  is the amplification factor such that  $E(k^2 v^2) = 1$ . Heuristically, this result can be understood in terms of our knowledge of (A) where  $\dim x = 1 = \dim u$ . Since the innovation term is independent of the first received signal  $z_{21}$ , the sender, in sending the second signal, essentially faces a new (A) type of problem which is known to possess linear solutions. Roughly speaking, we have transformed via noiseless feedback a problem of unequal source and channel rates to that of equal rates.

#### IV. ZERO- AND NONZERO-SUM VERSIONS OF SIGNALING

While none of the results in the previous sections taken by themselves are particularly significant, taken together they do provide considerable insight into the relationship between nonclassical decision and control theory on the one hand and Shannon's information theory on the other. We see how knowledge in one subject provides solutions in another. In fact, results in (A-1)–(A-5) form a significant portion of *all* the nontrivial knowledge concerning explicit solutions to dynamic information structure problems. Information-theoretic results play a crucial role in the solution or partial solution of these problems. On the other hand, viewed in this light, we also realize that the information theory problem is a very special kind of problem in dynamic teams. In a sense, it is the simplest kind of such problem: only two  $DM$ 's are involved and are explicitly and exclusively concerned with signaling. As we have mentioned briefly in Section II, a natural generalization to other classes of dynamic information structure problem exist. This development will be pursued now.

Once we consider the nonzero- or zero-sum version of

<sup>6</sup>See the Appendix for examples of (A-3) and (A-4) where explicit nonlinear schemes are illustrated and which are better than the best linear schemes.

the signaling problem, (P-2)' and (P-3) become the governing conditions of optimality. This permits a considerable simplification. Either one of the two inequalities of (P-2)' of (P-3) defines a *one-person* decision problem for *fixed* strategy of the other *DM*. The difficulties of solution-dependent convexity discussed in Section II are largely ameliorated. One still has to solve a pair of implicit equations in  $(\gamma_1, \gamma_2)$ . But this is a much simpler task as the discussion below will show.

A current problem of interest in economics is that of market signaling by Spence [14]. In terms of our basic formulation, the problem can be stated as follows. An employer must hire someone for a job without knowing how productive that individual will be. In other words, the employer has imperfect information about an individual's ability. Spence suggests that the employer can improve his information by looking on the job application for some signal, such as educational level. The employer offers wages based on the signal he sees; that is, a person with more education is offered higher wages, because the employer believes that the higher education indicates higher ability. The individual applying for the job, on the other hand, knowing he will receive wages based on his educational level, must decide how much education to get, taking into consideration that education is costly. Let  $DM_1$  = all potential employees considered together,  $DM_2$  = the employer,  $x$  = an individual's ability (known to that individual, but not to the employer),  $u_1$  (or  $u_1 + \text{noise}$ ) = educational level, and  $u_2$  = wages. The payoff or loss function of  $DM_2$  is  $[x - u_2]^2$ ; he does not wish to overpay or underpay with respect to  $x$ . The payoff of  $DM_1$  is simply the net profit  $u_2 - c(u_1, x)$  where  $c$  is the cost of signaling. Thus, we have precisely the following example, where the appropriate optimality conditions are (P-2)'

$$(B) \begin{cases} \max J_1 = E[\gamma_2(z_2) - c(\gamma_1(z_1), x)] \\ \min J_2 = E[(\gamma_2(z_2) - x)^2] \\ \text{where } z_1 = x \\ \quad \quad z_2 = u_1 \text{ or } u_1 + v \\ p(x, v) \text{ given.} \end{cases}$$

A reasonable special case of (B) is for  $c(u_1, x) = u_1/x$ ,  $p(x, v) = p(x)p(v)$  each being a uniform distribution,  $U_1 =$  a discrete set, and  $U_2 = R^+$ . Under this and other similar set-ups, equilibrium solutions  $\gamma_1^*, \gamma_2^*$  can be obtained. The details and economic interpretations are available elsewhere [3], [21]. Two noteworthy features of the solution are worth mentioning.

i) There are multiple equilibrium solutions  $(\gamma_1^*, \gamma_2^*)$ . This is a phenomenon that seems to occur only with dynamic information structure. Essentially, the equilibrium conditions are not sufficiently constraining, so that a large number of  $(\gamma_1, \gamma_2)$  pairs can satisfy them. It is for the same reason that team solutions satisfying (P-2) do not usually produce solutions which also satisfy (P-1) in the case  $J_1 = J_2$ . (P-2) is far from sufficient a condition. On the other hand, in static team problems, (P-2), under

reasonable conditions on  $J$ , often turns out to be necessary *and* sufficient [2].

ii) There are threshold phenomena in market signaling. If the cost of signaling is too high, or the signaling channel too noisy, or the underlying signal  $x$  itself too predictable, then signaling will suddenly cease altogether, i.e.,  $u_1 = 0$ . This phenomenon may be due to the nonzero-sum nature of the problem. In the cooperative case of information theory, it is always worthwhile to send some message, at least in the Gaussian case.

Finally, we can consider the case of  $J_1 = -J_2$ . As described earlier, we have a situation of "antisingaling". A prototype problem can be formulated by slight modification of (6)

$$(C) \begin{cases} \text{Find the saddle point pair } (\gamma_1^*, \gamma_2^*) \text{ for} \\ L(u, \xi) = \frac{1}{2}(x + au_1 + bu_2)^2 + \frac{c}{2}u_1^2 - \frac{d}{2}u_2^2 \\ z_1 = x \\ z_2 = gx + hu_1 + v \\ a, b, g, h \geq 0 \quad c, d > 0 \\ x \sim N(0, 1) \quad v \sim N(0, \sigma^2), \quad x, v \text{ independent.} \end{cases}$$

(C) is different from (2) only in the addition of the  $-d/2u_2^2$  term in  $L(u, \xi)$  and maximization with respect to  $u_2 = \gamma_2(z_2)$  instead of minimization. In addition to the advantage of solving only for equilibrium solutions, we have the added structure of  $J_1 = -J_2$ . Any saddle point solution is as good globally as any other solution on the product set of admissible  $(\gamma_1, \gamma_2)$  solutions by virtue of interchangability [15, p. 66]. Linear or affine saddle point strategies can be obtained for (C). In fact, (C) can be generalized considerably to include state (as well as information) dynamics resulting in a stochastic differential game problem and solved similarly [16], [17]. Other non LQG setups are also possible [18], [19]. The underlying idea of solutions is apparently a tradeoff between "revealing knowledge of  $x$  through  $u_1$ " versus "achieving some desirable payoff through  $u_1$ ".

In terms of information-theoretic ideas, a possible further tie with discussion in this paper is through the problem of wiretap channel [20] which clearly embodies the concept of "anti-signaling". However, we shall leave the formulation and unification of these ideas to future work of interested parties.

## V. CONCLUSION

The previous discussions can be summarized in Fig. 1. Considerable obvious and easy generalizations of the results to (A-1)-(A-5) are possible. However, nothing conceptually new is added.

Several conclusions can be drawn from this study.

i) Simple two-person decision problems with dynamic information structure have many interesting areas of ap-

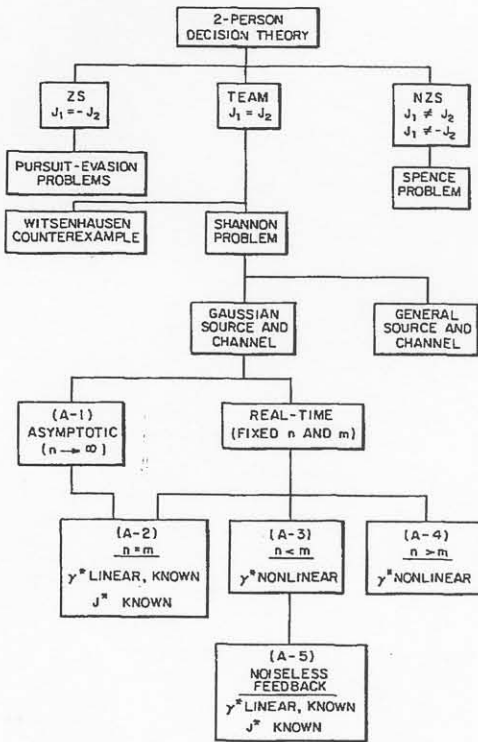


Fig. 1. Teams, signaling, and information theory.

plication, such as real-time (fixed block length) encoder and decoder design, economic theory, and (possibly) cryptography.

ii) Dynamic information structure leads to a new kind of deterministic optimization problem in which composition of functions is involved, namely,  $J(\gamma_1, \gamma_2(\gamma_1))$ . No reasonable algorithm seems to exist for this class of problems.

APPENDIX<sup>7</sup>

A. (A-3): One Sample to Two Signals Encoding and Decoding

$$x \sim N(0, 1) \quad v \sim N(0, \sigma^2 I_2).$$

Divide  $x$  into four equiprobable regions, as shown in Fig. 2. For the encoder, let  $u_{11}$  represent the region  $r(x)$  that  $x$  is from and  $u_{12}$  be a linear transformation of  $x$  in a stretched out version of this region (see Fig. 2). More precisely,  $u_{11} = cr(x)$  and

$$u_{12} = B \left( \frac{2x}{A} + 5 - 2r(x) \right),$$

where  $c$  and  $B$  are chosen so that the power constraint is

<sup>7</sup>See Appendices III-B and III-C in [21] for details.

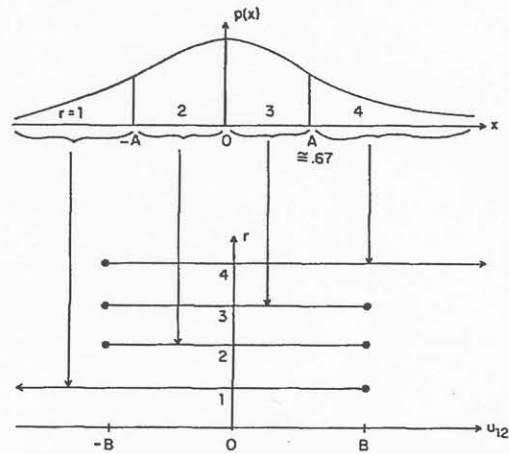


Fig. 2. Stretched mapping of  $x$  to two dimensions.

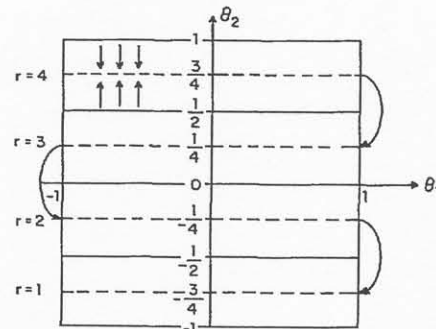


Fig. 3. Transformation of square to dotted line.

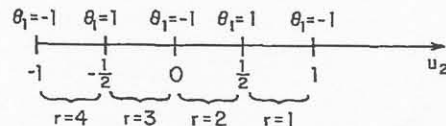


Fig. 4. Transformation of dotted line to  $u_3$ .

satisfied. Let the estimates be

$$\hat{r} = \arg \max_r p(r/z_{21}), \quad z_{21} = cr + v_1.$$

$$\hat{u}_{12} = z_{22} = u_{12} + v_2.$$

Then, inverting the expression for  $u_{12}$ , yields

$$\hat{x} = \frac{A}{2} \left( \frac{\hat{u}_{12}}{B} - 5 + 2\hat{r} \right).$$

For small noise, the mean square error distortion with this scheme is better than with any linear scheme.

B. (A-4): Two Samples to One Signal Encoding and Decoding

$$x \sim N(0, I_2) \quad v \sim N(0, \sigma^2).$$

Transform  $x_i$  to  $\theta_i = 2/\pi \tan^{-1} x_i, i = 1, 2$ . Then  $\theta_i \in [-1, 1]$ .

Map all points in the resulting square to the dotted lines  $r = 1, 2, 3, 4$ , as shown in Fig. 3, where

$$r(\theta_2) = \begin{cases} 1 & -1 \leq \theta_2 \leq -\frac{1}{2} \\ 2 & -\frac{1}{2} < \theta_2 \leq 0 \\ 3 & 0 < \theta_2 \leq \frac{1}{2} \\ 4 & \frac{1}{2} < \theta_2 \leq 1. \end{cases}$$

Straighten out the dotted line and compress it to fit into the interval  $[-1, 1]$ , and call the variable  $u_1$ , as shown in Fig. 4. Then it can be shown that

$$u_1 = \frac{1}{4} [(-1)^r \theta_1 + 5 - 2r].$$

Let the estimates be  $\hat{u}_1 = z_1 = u_1 + v$  and

$$\hat{r} = \begin{cases} 1 & \text{if } \frac{1}{2} < \hat{u}_1 \\ 2 & \text{if } 0 < \hat{u}_1 \leq \frac{1}{2} \\ 3 & \text{if } -\frac{1}{2} < \hat{u}_1 \leq 0 \\ 4 & \text{if } \hat{u}_1 \leq -\frac{1}{2} \end{cases}$$

$$\hat{\theta}_1 = \begin{cases} (-1)^{\hat{r}} (4\hat{u}_1 - 5 + 2\hat{r}), & -1 \leq \hat{u}_1 \leq 1 \\ -1, & \hat{u}_1 < -1 \\ 1, & \hat{u}_1 > 1 \end{cases}$$

so that  $\hat{\theta}_1 \in [-1, 1]$ . Then let  $\hat{\theta}_2 = (5 - 2\hat{r})/4$  and  $\hat{x}_i = \pi/2 \tan \hat{\theta}_i, i = 1, 2$ . Again, for small noise, the mean square error distortion with this scheme is better than with any linear scheme.

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# The Optimal Decentralized Control of a Large Power System: Load and Frequency Control

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**Abstract**—The load and frequency control of a multi-area interconnected power system is studied. In this problem, the system is assumed to be subject to unknown constant disturbances, and it is desired to obtain, if possible, robust decentralized controllers so that the frequency and tie-line/net-area power flow of the power system are regulated. The problem is solved by using some structural results recently obtained in decentralized control, in conjunction with a parameter optimization method which minimizes the dominant eigenvalue of the closed-loop system. A class of minimum order robust decentralized controllers which solves this general multi-area load and frequency control problem is obtained. Application of these results is then made to solve the load and frequency control problem for a power system consisting of nine synchronous machines (described by a 119th-order system). It is shown that the load and frequency controller obtained in this case is not likely to be significantly improved by using more complex controllers; in particular, it is shown that the conventional controller, used in regulating the net-area power flow of a system, is not likely to be significantly improved upon by using more complex controllers.

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## I. INTRODUCTION

THE purpose of this paper is to use some structural results recently obtained in decentralized control to study the load and frequency control problem for a multi-area interconnected power system, and then obtain, if possible, realistic minimum order controllers which will solve the problem. Elgerd and Fosha [1], [2] studied this problem for a two-area system based on a ninth-order model, without explicitly dealing with the decentralization constraint. Since this study, additional work [3]-[6] based on 17th-order models or less has been carried out for the problem, again without explicitly dealing with the decentralization constraint. In this work, a minimum order robust decentralized controller will be proposed to solve this general multi-area load and frequency control problem, and it will be shown that the controller is not likely to be significantly improved using higher order compensators. The main contribution of the proposed design method over existing design methods is that 1) the decentralized controller constraint is explicitly taken into